

# 具有热保护功能的 TPS22810-Q1 2.7V-18V、79mΩ 导通电阻负载开关

## 1 特性

- 符合汽车类标准
- 具有符合 AEC-Q100 标准的下列特性：
  - 器件温度等级 2：环境工作温度范围为  $-40^{\circ}\text{C}$  至  $+105^{\circ}\text{C}$
  - 器件 HBM ESD 分类等级 2
  - 器件 CDM ESD 分类等级 C5
- 集成单通道负载开关
- 最大连续电流为 2A
- 输入电压：2.7V 至 18V
- 绝对最大输入电压：20V
- 导通电阻 ( $R_{\text{ON}}$ )
  - $R_{\text{ON}} = 79\text{m}\Omega$  ( $V_{\text{IN}} = 12\text{V}$  时的典型值)
- 静态电流
  - $62\mu\text{A}$  ( $V_{\text{IN}} = 12\text{V}$  时的典型值)
- 关断电流
  - $500\text{nA}$  ( $V_{\text{IN}} = 12\text{V}$  时的典型值)
- 热关断
- 欠压闭锁 (UVLO)
- 可调节快速输出放电 (QOD)
- 可通过 CT 引脚配置的上升时间
- 小外形尺寸晶体管 (SOT) 23-6 封装
  - $2.9\text{mm} \times 2.8\text{mm}$ , 0.95mm 间距, 1.45mm 高 (DBV)

## 2 应用

- 汽车音响主机
- 环视 ECU

## 3 说明

TPS22810-Q1 是一款单通道负载开关，具有可配置的上升时间和集成式快速输出放电 (QOD) 功能。该器件具有热关断功能，可保护器件免受高结温的损坏，因此可从根本上确保器件的安全运行区。该器件具有一个 N 沟道 MOSFET，可在 2.7V 至 18V 的输入电压范围内运行。该器件可支持 2A 的最大电流。开关可由一个打开和关闭输入控制，此输入可直接连接至低压控制信号。

该器件的可配置上升时间可大幅降低大容量负载电容所产生的浪涌电流，从而降低或消除电源压降。欠压闭锁用于在输入电压降至阈值以下时关闭器件，以确保下游电路不会因为供电电压低于预期值而损坏。可配置的快速输出放电 (QOD) 引脚控制器件的下降时间，以便针对掉电进行灵活设计。

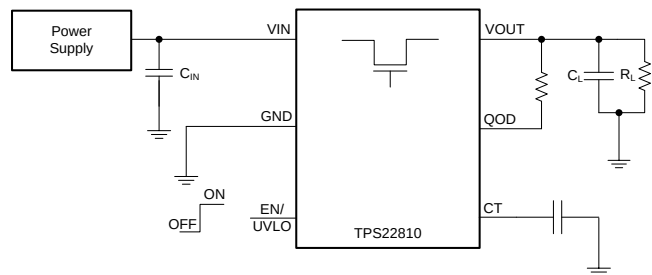
TPS22810-Q1 器件可提供方便目测检查焊点的带引线的 SOT-23 封装 (DBV)。该器件在自然通风环境下的额定运行温度范围为  $-40^{\circ}\text{C}$  至  $+105^{\circ}\text{C}$ 。

器件信息<sup>(1)</sup>

| 器件型号        | 封装         | 封装尺寸 (标称值)      |
|-------------|------------|-----------------|
| TPS22810-Q1 | SOT-23 (6) | 2.90mm x 2.80mm |

(1) 如需了解所有可用封装，请参阅数据表末尾的可订购产品附录。

简化原理图



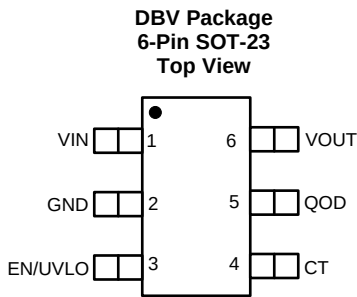
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## 4 修订历史记录

| 日期      | 修订版本 | 说明      |
|---------|------|---------|
| 4 月2018 | *    | 最初发布版本。 |

## 5 Pin Configuration and Functions



### Pin Functions

| PIN     |     | I/O | DESCRIPTION                                                                                                                                                                                                                                                                                                                                                                                                      |
|---------|-----|-----|------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------|
| NAME    | NO. |     |                                                                                                                                                                                                                                                                                                                                                                                                                  |
| CT      | 4   | O   | Switch slew rate control. Can be left floating                                                                                                                                                                                                                                                                                                                                                                   |
| EN/UVLO | 3   | I   | Active high switch control input and UVLO adjustment. Do not leave floating                                                                                                                                                                                                                                                                                                                                      |
| GND     | 2   | —   | Device ground                                                                                                                                                                                                                                                                                                                                                                                                    |
| QOD     | 5   | O   | Quick Output Discharge pin. This functionality can be enabled in one of three ways: <ul style="list-style-type: none"> <li>Placing an external resistor between VOUT and QOD</li> <li>Tying QOD directly to VOUT and using the internal resistor value (<math>R_{PD}</math>)</li> <li>Disabling QOD by leaving pin floating</li> </ul> See the <a href="#">Quick Output Discharge (QOD)</a> for more information |
| VIN     | 1   | I   | Switch input. Place ceramic bypass capacitor(s) between this pin and GND                                                                                                                                                                                                                                                                                                                                         |
| VOUT    | 6   | O   | Switch output                                                                                                                                                                                                                                                                                                                                                                                                    |

## 6 Specifications

### 6.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)<sup>(1)</sup>

|                      |                                  |                  | MIN  | MAX                              | UNIT |
|----------------------|----------------------------------|------------------|------|----------------------------------|------|
| V <sub>IN</sub>      | Maximum Input Voltage Range      | V <sub>IN</sub>  | –0.3 | 20                               | V    |
| V <sub>OUT</sub>     | Maximum Output Voltage Range     | V <sub>OUT</sub> | –0.3 | min (20V, V <sub>IN</sub> + 0.3) |      |
| V <sub>EN/UVLO</sub> | Maximum Enable Pin Voltage Range | EN/UVLO          | –0.3 | 20                               | V    |
| T <sub>J</sub>       | Junction temperature             |                  |      | 150                              | °C   |
| T <sub>stg</sub>     | Storage temperature              |                  | –65  | 150                              | °C   |

- (1) Stresses beyond those listed under *Absolute Maximum Rating* may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Condition*. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

### 6.2 ESD Ratings

|                    |                         | VALUE                                                   | UNIT  |
|--------------------|-------------------------|---------------------------------------------------------|-------|
| V <sub>(ESD)</sub> | Electrostatic discharge | Human body model (HBM), per AEC Q100-002 <sup>(1)</sup> | ±3000 |
|                    |                         | Charged device model (CDM), per AEC Q100-011            | ±750  |
|                    |                         | Other pins                                              | ±1000 |

- (1) AEC Q100-002 indicates that HBM stressing shall be in accordance with the ANSI/ESDA/JEDEC JS-001 specification.

### 6.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

|                      |                                               |           | MIN | MAX             | UNIT |
|----------------------|-----------------------------------------------|-----------|-----|-----------------|------|
| V <sub>IN</sub>      | Input Voltage Range                           | IN        | 2.7 | 18              | V    |
| V <sub>OUT</sub>     | Output Voltage Range                          | OUT       |     | V <sub>IN</sub> | V    |
| V <sub>EN/UVLO</sub> | Enable Pin Voltage Range                      | EN/UVLO   | 0   | 18              | V    |
| I <sub>MAX</sub>     | Maximum continuous switch current, TA = 65°C  | IN to OUT |     | 2               | A    |
| I <sub>MAX</sub>     | Maximum continuous switch current, TA = 85°C  | IN to OUT |     | 1.5             | A    |
| I <sub>MAX</sub>     | Maximum continuous switch current, TA = 105°C | IN to OUT |     | 1               | A    |
| T <sub>A</sub>       | Operating free-air temperature                |           | –40 | 105             | °C   |
| C <sub>IN</sub>      | Input Capacitor <sup>(1)</sup>                |           | 1   |                 | μF   |

- (1) See the Detailed Description section.

### 6.4 Thermal Information

| THERMAL METRIC <sup>(1)</sup> |                                              | TPS22810-Q1 | UNIT |
|-------------------------------|----------------------------------------------|-------------|------|
|                               |                                              | DBV (SOT23) |      |
|                               |                                              | 6 PINS      |      |
| R <sub>θJA</sub>              | Junction-to-ambient thermal resistance       | 182         | °C/W |
| R <sub>θJC(top)</sub>         | Junction-to-case (top) thermal resistance    | 127.2       | °C/W |
| R <sub>θJB</sub>              | Junction-to-board thermal resistance         | 16.9        | °C/W |
| Ψ <sub>JT</sub>               | Junction-to-top characterization parameter   | 26.4        | °C/W |
| Ψ <sub>JB</sub>               | Junction-to-board characterization parameter | 36.3        | °C/W |

- (1) For more information about traditional and new thermal metrics, see the [Semiconductor and IC Package Thermal Metrics](#) application report.

## 6.5 Electrical Characteristics

Unless otherwise noted, the specification in the following table applies over the following ambient operating temperature  $-40^{\circ}\text{C} \leq T_A \leq +105^{\circ}\text{C}$ . Typical values are for  $T_A = 25^{\circ}\text{C}$ .

| PARAMETER     |                                             | TEST CONDITIONS                                   |                                                      | MIN                                                  | TYP  | MAX  | UNIT          |
|---------------|---------------------------------------------|---------------------------------------------------|------------------------------------------------------|------------------------------------------------------|------|------|---------------|
| $I_{Q, VIN}$  | Quiescent current                           | $I_{OUT} = 0\text{ A}$                            | $V_{IN} = 18\text{ V}$                               | $T_A = -40^{\circ}\text{C to } +85^{\circ}\text{C}$  |      | 62   | 80            |
|               |                                             |                                                   |                                                      | $T_A = -40^{\circ}\text{C to } +105^{\circ}\text{C}$ |      |      | 85            |
|               |                                             |                                                   | $V_{IN} = 12\text{ V}$                               | $T_A = -40^{\circ}\text{C to } +85^{\circ}\text{C}$  |      | 62   | 80            |
|               |                                             |                                                   |                                                      | $T_A = -40^{\circ}\text{C to } +105^{\circ}\text{C}$ |      |      | 85            |
|               |                                             |                                                   | $V_{IN} = 5\text{ V}$                                | $T_A = -40^{\circ}\text{C to } +85^{\circ}\text{C}$  |      | 59   | 80            |
|               |                                             |                                                   |                                                      | $T_A = -40^{\circ}\text{C to } +105^{\circ}\text{C}$ |      |      | 85            |
|               |                                             |                                                   | $V_{IN} = 3.3\text{ V}$                              | $T_A = -40^{\circ}\text{C to } +85^{\circ}\text{C}$  |      | 53   | 80            |
|               |                                             |                                                   |                                                      | $T_A = -40^{\circ}\text{C to } +105^{\circ}\text{C}$ |      |      | 85            |
|               |                                             |                                                   | $V_{IN} = 2.7\text{ V}$                              | $T_A = -40^{\circ}\text{C to } +85^{\circ}\text{C}$  |      | 49   | 70            |
|               |                                             |                                                   |                                                      | $T_A = -40^{\circ}\text{C to } +105^{\circ}\text{C}$ |      |      | 85            |
| $I_{SD, VIN}$ | Shutdown current                            | $V_{EN} = 0\text{ V}, V_{OUT} = 0\text{ V}$       | $V_{IN} = 18\text{ V}$                               | $T_A = -40^{\circ}\text{C to } +85^{\circ}\text{C}$  |      | 0.5  | 2.3           |
|               |                                             |                                                   |                                                      | $T_A = -40^{\circ}\text{C to } +105^{\circ}\text{C}$ |      |      | 3.8           |
|               |                                             |                                                   | $V_{IN} = 12\text{ V}$                               | $T_A = -40^{\circ}\text{C to } +85^{\circ}\text{C}$  |      | 0.5  | 2.3           |
|               |                                             |                                                   |                                                      | $T_A = -40^{\circ}\text{C to } +105^{\circ}\text{C}$ |      |      | 3.8           |
|               |                                             |                                                   | $V_{IN} = 5\text{ V}$                                | $T_A = -40^{\circ}\text{C to } +85^{\circ}\text{C}$  |      | 0.5  | 2.3           |
|               |                                             |                                                   |                                                      | $T_A = -40^{\circ}\text{C to } +105^{\circ}\text{C}$ |      |      | 3.8           |
|               |                                             |                                                   | $V_{IN} = 3.3\text{ V}$                              | $T_A = -40^{\circ}\text{C to } +85^{\circ}\text{C}$  |      | 0.5  | 2.3           |
|               |                                             |                                                   |                                                      | $T_A = -40^{\circ}\text{C to } +105^{\circ}\text{C}$ |      |      | 3.8           |
|               |                                             |                                                   | $V_{IN} = 2.7\text{ V}$                              | $T_A = -40^{\circ}\text{C to } +85^{\circ}\text{C}$  |      | 0.5  | 2.3           |
|               |                                             |                                                   |                                                      | $T_A = -40^{\circ}\text{C to } +105^{\circ}\text{C}$ |      |      | 3.8           |
| $I_{EN/UVLO}$ | EN/UVLO pin input leakage current           | $V_{IN} = 18\text{ V}, I_{OUT} = 0\text{ A}$      | $T_A = -40^{\circ}\text{C to } +105^{\circ}\text{C}$ |                                                      |      | 0.1  | $\mu\text{A}$ |
| $V_{UVR}$     | VIN UVLO threshold, rising                  |                                                   | $T_A = -40^{\circ}\text{C to } +105^{\circ}\text{C}$ |                                                      | 2    | 2.54 | 2.62          |
| $V_{UVRhyst}$ | VIN UVLO hysteresis                         |                                                   | $T_A = -40^{\circ}\text{C to } +105^{\circ}\text{C}$ |                                                      |      | 5    | %             |
| $V_{ENR}$     | EN threshold, rising                        |                                                   | $T_A = -40^{\circ}\text{C to } +105^{\circ}\text{C}$ |                                                      | 1.13 | 1.23 | 1.3           |
| $V_{ENF}$     | EN threshold, falling                       |                                                   | $T_A = -40^{\circ}\text{C to } +105^{\circ}\text{C}$ |                                                      | 1.08 | 1.13 | 1.18          |
| $V_{SHUTF}$   | EN threshold voltage for low $I_Q$ shutdown |                                                   | $T_A = -40^{\circ}\text{C to } +105^{\circ}\text{C}$ |                                                      | 0.5  | 0.75 | 0.9           |
| $R_{ON}$      | On-resistance                               | $V_{IN} = 18\text{ V}, I_{OUT} = -200\text{ mA}$  | $T_A = 25^{\circ}\text{C}$                           |                                                      |      | 79   | 86            |
|               |                                             |                                                   | $T_A = -40^{\circ}\text{C to } +85^{\circ}\text{C}$  |                                                      |      |      | 105           |
|               |                                             |                                                   | $T_A = -40^{\circ}\text{C to } +105^{\circ}\text{C}$ |                                                      |      |      | 115           |
|               |                                             | $V_{IN} = 12\text{ V}, I_{OUT} = -200\text{ mA}$  | $T_A = 25^{\circ}\text{C}$                           |                                                      |      | 79   | 86            |
|               |                                             |                                                   | $T_A = -40^{\circ}\text{C to } +85^{\circ}\text{C}$  |                                                      |      |      | 105           |
|               |                                             |                                                   | $T_A = -40^{\circ}\text{C to } +105^{\circ}\text{C}$ |                                                      |      |      | 115           |
|               |                                             | $V_{IN} = 9\text{ V}, I_{OUT} = -200\text{ mA}$   | $T_A = 25^{\circ}\text{C}$                           |                                                      |      | 79   | 86            |
|               |                                             |                                                   | $T_A = -40^{\circ}\text{C to } +85^{\circ}\text{C}$  |                                                      |      |      | 105           |
|               |                                             |                                                   | $T_A = -40^{\circ}\text{C to } +105^{\circ}\text{C}$ |                                                      |      |      | 115           |
|               |                                             | $V_{IN} = 5\text{ V}, I_{OUT} = -200\text{ mA}$   | $T_A = 25^{\circ}\text{C}$                           |                                                      |      | 79   | 86            |
|               |                                             |                                                   | $T_A = -40^{\circ}\text{C to } +85^{\circ}\text{C}$  |                                                      |      |      | 105           |
|               |                                             |                                                   | $T_A = -40^{\circ}\text{C to } +105^{\circ}\text{C}$ |                                                      |      |      | 115           |
|               |                                             | $V_{IN} = 3.3\text{ V}, I_{OUT} = -200\text{ mA}$ | $T_A = 25^{\circ}\text{C}$                           |                                                      |      | 83   | 92            |
|               |                                             |                                                   | $T_A = -40^{\circ}\text{C to } +85^{\circ}\text{C}$  |                                                      |      |      | 115           |
|               |                                             |                                                   | $T_A = -40^{\circ}\text{C to } +105^{\circ}\text{C}$ |                                                      |      |      | 125           |
|               |                                             | $V_{IN} = 2.7\text{ V}, I_{OUT} = -200\text{ mA}$ | $T_A = 25^{\circ}\text{C}$                           |                                                      |      | 86   | 95            |
|               |                                             |                                                   | $T_A = -40^{\circ}\text{C to } +85^{\circ}\text{C}$  |                                                      |      |      | 120           |
|               |                                             |                                                   | $T_A = -40^{\circ}\text{C to } +105^{\circ}\text{C}$ |                                                      |      |      | 130           |

## Electrical Characteristics (continued)

Unless otherwise noted, the specification in the following table applies over the following ambient operating temperature  $-40^{\circ}\text{C} \leq T_A \leq +105^{\circ}\text{C}$ . Typical values are for  $T_A = 25^{\circ}\text{C}$ .

| PARAMETER                                | TEST CONDITIONS                                               | MIN                                                   | TYP | MAX | UNIT               |
|------------------------------------------|---------------------------------------------------------------|-------------------------------------------------------|-----|-----|--------------------|
| $R_{PD}$ Output pull down resistance     | $V_{IN} = V_{OUT} = 18\text{ V}$ , $V_{EN/UVLO} = 0\text{ V}$ | $T_A = -40^{\circ}\text{C}$ to $+105^{\circ}\text{C}$ | 290 | 350 | $\Omega$           |
|                                          | $V_{IN} = V_{OUT} = 12\text{ V}$ , $V_{EN/UVLO} = 0\text{ V}$ | $T_A = -40^{\circ}\text{C}$ to $+105^{\circ}\text{C}$ | 265 | 350 |                    |
|                                          | $V_{IN} = V_{OUT} = 5\text{ V}$ , $V_{EN/UVLO} = 0\text{ V}$  | $T_A = -40^{\circ}\text{C}$ to $+105^{\circ}\text{C}$ | 250 | 400 |                    |
| $T_{SD}$ Thermal shutdown threshold      | $V_{IN} = 18\text{ V}$                                        | $T_A = -40^{\circ}\text{C}$ to $+105^{\circ}\text{C}$ | 160 |     | $^{\circ}\text{C}$ |
| $T_{SD,HYS}$ Thermal shutdown hysteresis | $V_{IN} = 18\text{ V}$                                        | $T_A = -40^{\circ}\text{C}$ to $+105^{\circ}\text{C}$ | 30  |     | $^{\circ}\text{C}$ |

## 6.6 Switching Characteristics

Refer to the timing test circuit in [Figure 16](#) (unless otherwise noted) for references to external components used for the test condition in the switching characteristics table. Switching characteristics shown below are only valid for the power-up sequence where  $V_{IN}$  is already in steady state condition before the  $EN/UVLO$  pin is asserted high.

| PARAMETER                                                                                            |                            | TEST CONDITIONS                                                                                  | MIN | TYP | MAX | UNIT |
|------------------------------------------------------------------------------------------------------|----------------------------|--------------------------------------------------------------------------------------------------|-----|-----|-----|------|
| V <sub>IN</sub> = 18 V, V <sub>EN/UVLO</sub> = 5 V, T <sub>A</sub> = 25 °C (unless otherwise noted)  |                            |                                                                                                  |     |     |     |      |
| t <sub>ON</sub>                                                                                      | Turnon time                | R <sub>L</sub> = 10 Ω, C <sub>IN</sub> = 1 μF, C <sub>L</sub> = 0.1 μF, C <sub>T</sub> = 2200 pF |     | 520 |     | μs   |
| t <sub>OFF</sub>                                                                                     | Turnoff time               | R <sub>L</sub> = 10 Ω, C <sub>IN</sub> = 1 μF, C <sub>L</sub> = 0.1 μF, C <sub>T</sub> = 2200 pF |     | 3.3 |     |      |
| t <sub>R</sub>                                                                                       | V <sub>OUT</sub> rise time | R <sub>L</sub> = 10 Ω, C <sub>IN</sub> = 1 μF, C <sub>L</sub> = 0.1 μF, C <sub>T</sub> = 2200 pF |     | 700 |     |      |
| t <sub>F</sub>                                                                                       | V <sub>OUT</sub> fall time | R <sub>L</sub> = 10 Ω, C <sub>IN</sub> = 1 μF, C <sub>L</sub> = 0.1 μF, C <sub>T</sub> = 2200 pF |     | 2   |     |      |
| t <sub>D</sub>                                                                                       | Delay time                 | R <sub>L</sub> = 10 Ω, C <sub>IN</sub> = 1 μF, C <sub>L</sub> = 0.1 μF, C <sub>T</sub> = 2200 pF |     | 180 |     |      |
| V <sub>IN</sub> = 12 V, V <sub>EN/UVLO</sub> = 5 V, T <sub>A</sub> = 25 °C (unless otherwise noted)  |                            |                                                                                                  |     |     |     |      |
| t <sub>ON</sub>                                                                                      | Turnon time                | R <sub>L</sub> = 10 Ω, C <sub>IN</sub> = 1 μF, C <sub>L</sub> = 0.1 μF, C <sub>T</sub> = 2200 pF |     | 380 |     | μs   |
| t <sub>OFF</sub>                                                                                     | Turnoff time               | R <sub>L</sub> = 10 Ω, C <sub>IN</sub> = 1 μF, C <sub>L</sub> = 0.1 μF, C <sub>T</sub> = 2200 pF |     | 3.3 |     |      |
| t <sub>R</sub>                                                                                       | V <sub>OUT</sub> rise time | R <sub>L</sub> = 10 Ω, C <sub>IN</sub> = 1 μF, C <sub>L</sub> = 0.1 μF, C <sub>T</sub> = 2200 pF |     | 460 |     |      |
| t <sub>F</sub>                                                                                       | V <sub>OUT</sub> fall time | R <sub>L</sub> = 10 Ω, C <sub>IN</sub> = 1 μF, C <sub>L</sub> = 0.1 μF, C <sub>T</sub> = 2200 pF |     | 2   |     |      |
| t <sub>D</sub>                                                                                       | Delay time                 | R <sub>L</sub> = 10 Ω, C <sub>IN</sub> = 1 μF, C <sub>L</sub> = 0.1 μF, C <sub>T</sub> = 2200 pF |     | 150 |     |      |
| V <sub>IN</sub> = 3.3 V, V <sub>EN/UVLO</sub> = 5 V, T <sub>A</sub> = 25 °C (unless otherwise noted) |                            |                                                                                                  |     |     |     |      |
| t <sub>ON</sub>                                                                                      | Turnon time                | R <sub>L</sub> = 10 Ω, C <sub>IN</sub> = 1 μF, C <sub>L</sub> = 0.1 μF, C <sub>T</sub> = 2200 pF |     | 185 |     | μs   |
| t <sub>OFF</sub>                                                                                     | Turnoff time               | R <sub>L</sub> = 10 Ω, C <sub>IN</sub> = 1 μF, C <sub>L</sub> = 0.1 μF, C <sub>T</sub> = 2200 pF |     | 3.3 |     |      |
| t <sub>R</sub>                                                                                       | V <sub>OUT</sub> rise time | R <sub>L</sub> = 10 Ω, C <sub>IN</sub> = 1 μF, C <sub>L</sub> = 0.1 μF, C <sub>T</sub> = 2200 pF |     | 120 |     |      |
| t <sub>F</sub>                                                                                       | V <sub>OUT</sub> fall time | R <sub>L</sub> = 10 Ω, C <sub>IN</sub> = 1 μF, C <sub>L</sub> = 0.1 μF, C <sub>T</sub> = 2200 pF |     | 2   |     |      |
| t <sub>D</sub>                                                                                       | Delay time                 | R <sub>L</sub> = 10 Ω, C <sub>IN</sub> = 1 μF, C <sub>L</sub> = 0.1 μF, C <sub>T</sub> = 2200 pF |     | 130 |     |      |

## 6.7 Typical DC Characteristics

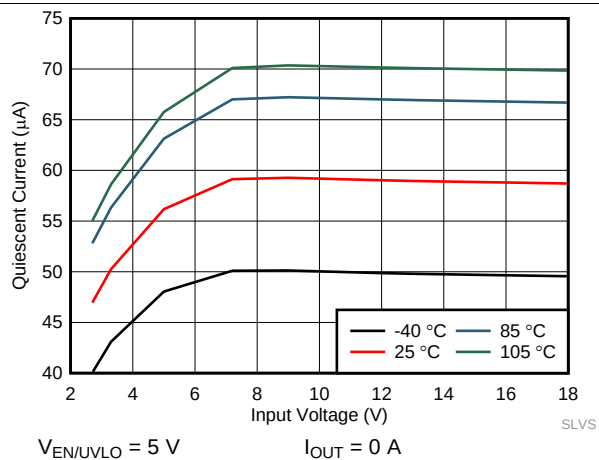


图 1. Quiescent Current vs Input Voltage

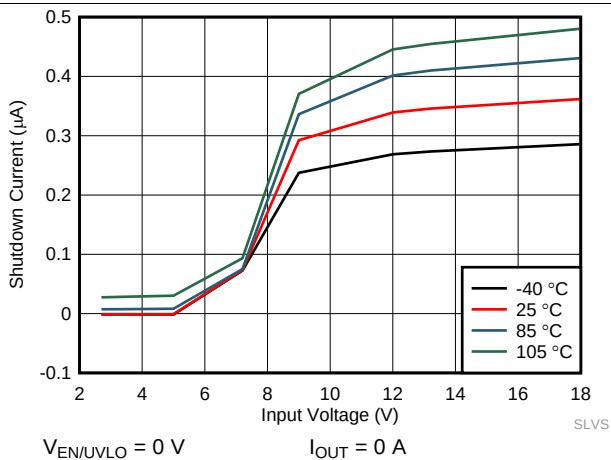


图 2. Shutdown Current vs Input Voltage

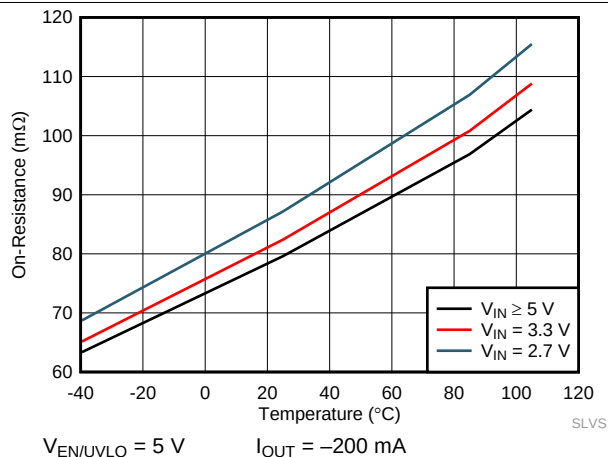


图 3. On-Resistance vs Temperature

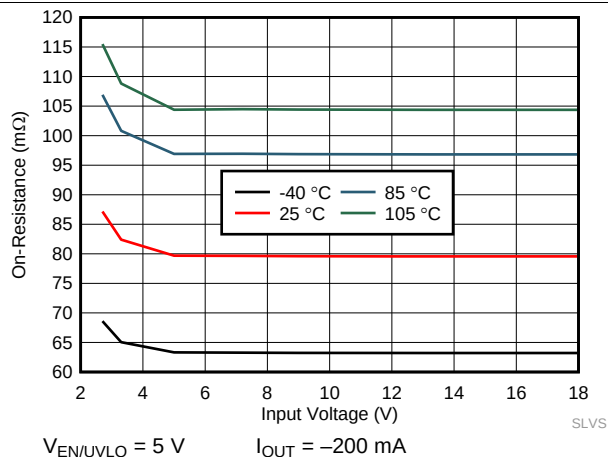


图 4. On-Resistance vs Input Voltage

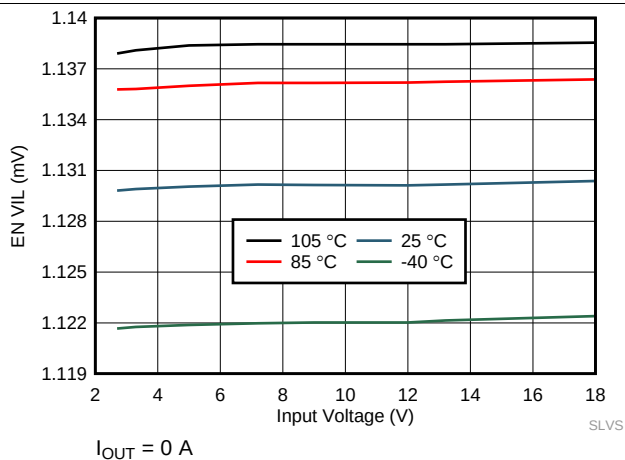


图 5. EN VIL vs Input Voltage

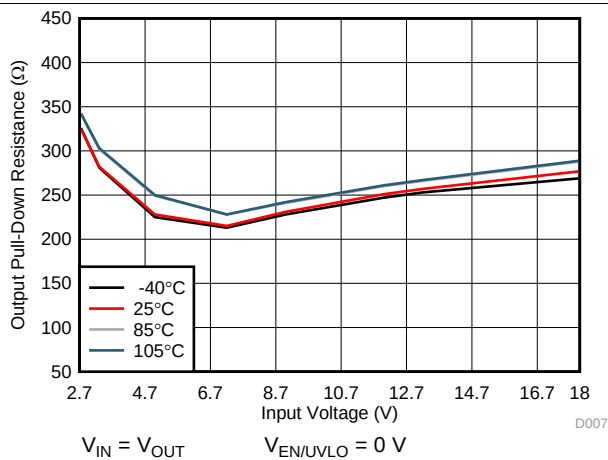


图 6. Output Pull-Down Resistance vs Input Voltage

## 6.8 Typical AC Characteristics

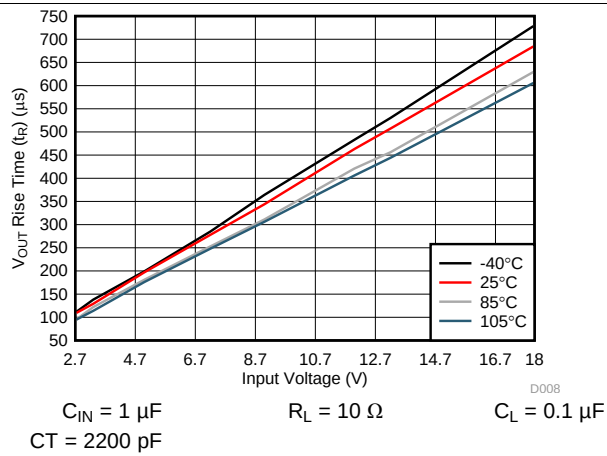


图 7.  $V_{OUT}$  Rise Time ( $t_R$ ) vs Input Voltage

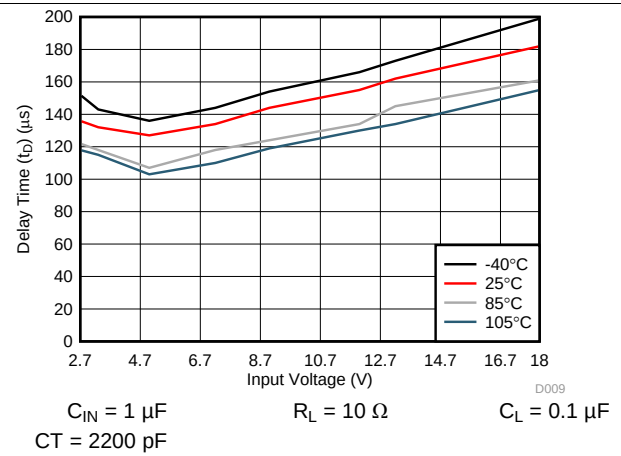


图 8. Delay Time ( $t_D$ ) vs Input Voltage

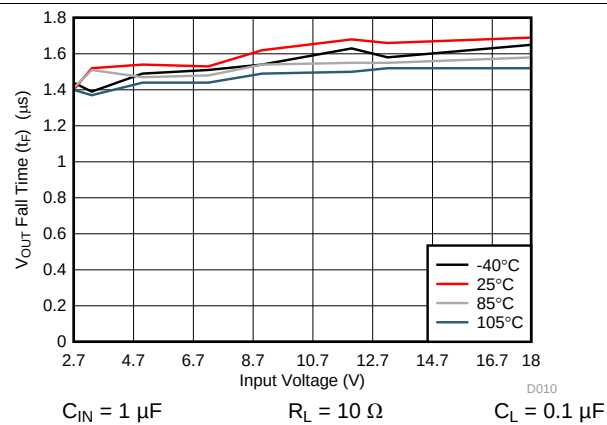


图 9.  $V_{OUT}$  Fall Time ( $t_F$ ) vs Input Voltage

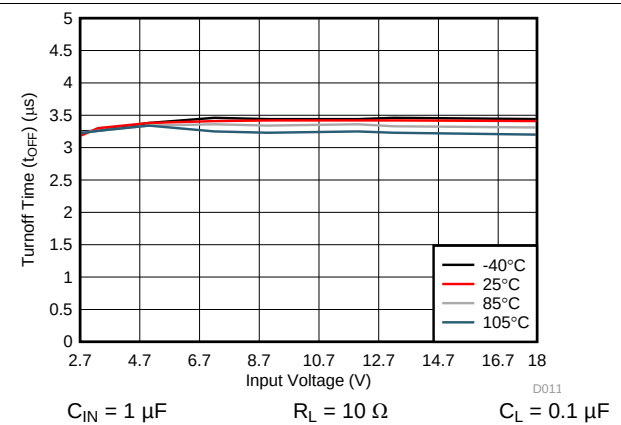


图 10. Turnoff Time ( $t_{OFF}$ ) vs Input Voltage

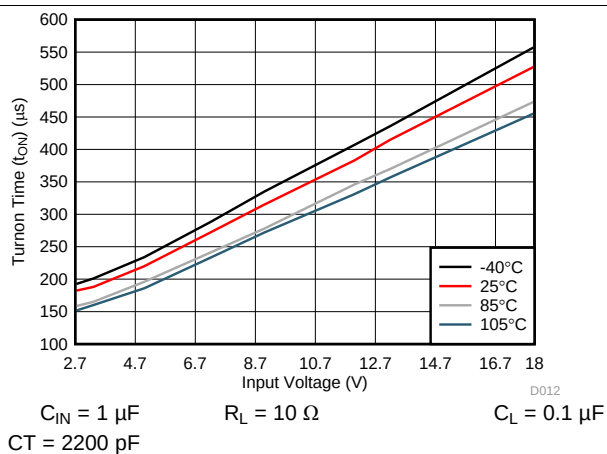


图 11. Turnon Time ( $t_{ON}$ ) vs Input Voltage

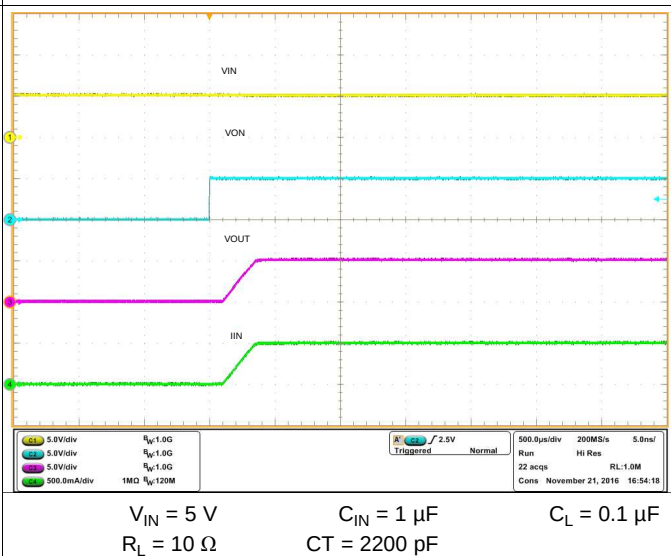
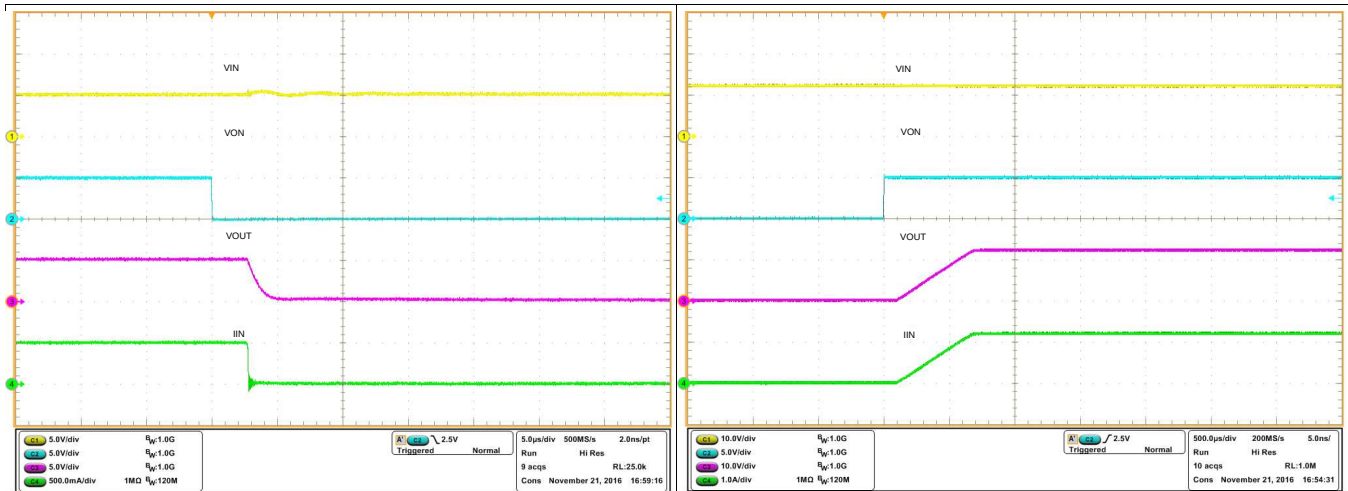


图 12. Rise Time  $t_R$  at  $V_{IN} = 5 V$



## Typical AC Characteristics (接下页)

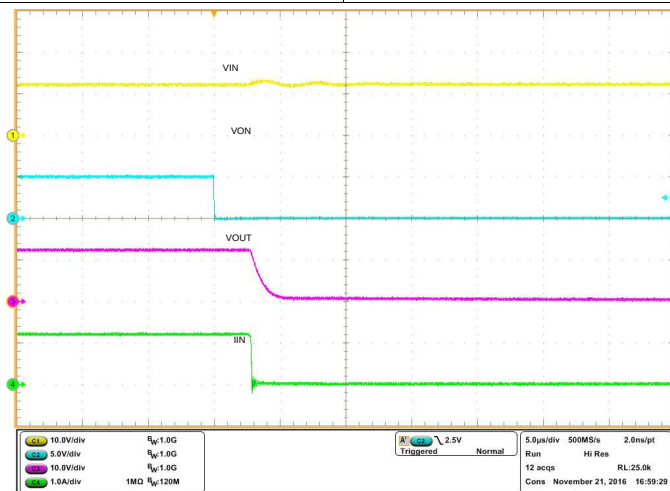


$V_{IN} = 5V$   
 $R_L = 10\Omega$   
 $C_{IN} = 1\mu F$   
 $QOD = \text{Open}$   
 $C_L = 0.1\mu F$

图 13. Fall Time  $t_F$  at  $V_{IN} = 5V$

$V_{IN} = 12V$   
 $R_L = 10\Omega$   
 $C_{IN} = 1\mu F$   
 $CT = 2200pF$   
 $C_L = 0.1\mu F$

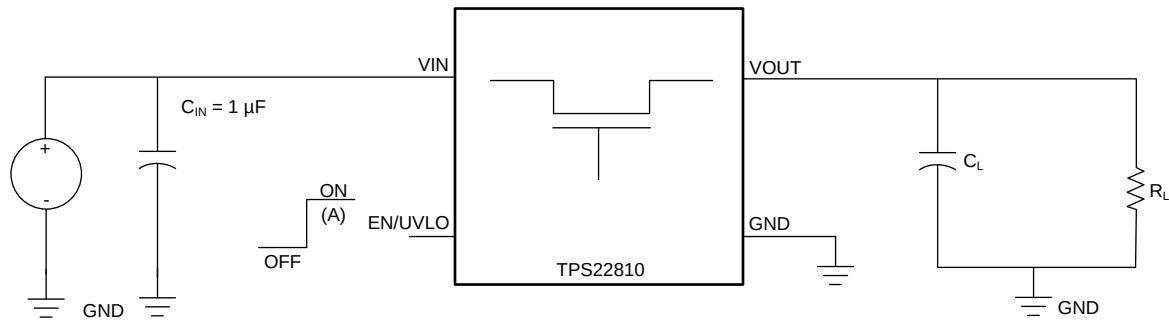
图 14. Rise Time  $t_R$  at  $V_{IN} = 12V$



$V_{IN} = 12V$   
 $R_L = 10\Omega$   
 $C_{IN} = 1\mu F$   
 $QOD = \text{Open}$   
 $C_L = 0.1\mu F$

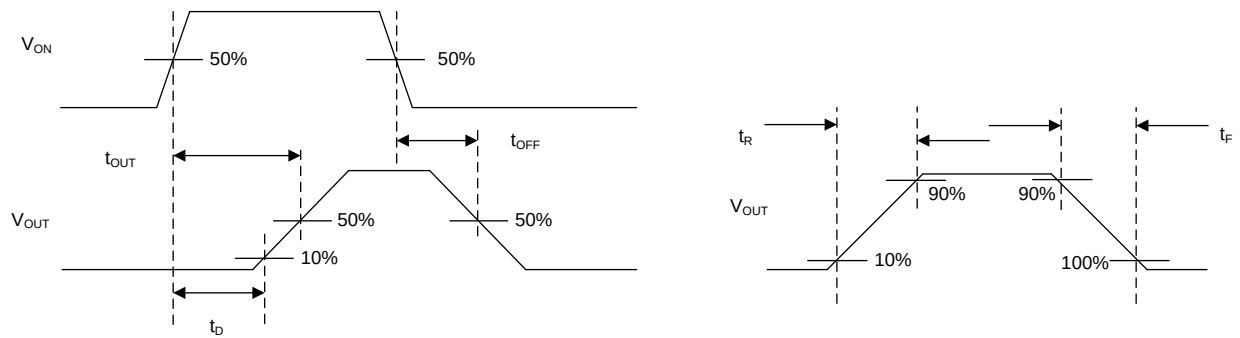
图 15. Fall Time  $t_F$  at  $V_{IN} = 12V$

## 7 Parameter Measurement Information



A. Rise and fall times of the control signal are 100 ns

**图 16. Test Circuit**



**图 17. Timing Waveforms**

## 8 Detailed Description

### 8.1 Overview

The TPS22810-Q1 is a 6-pin, 2.7-18-V load switch with thermal protection. To reduce voltage drop for low voltage and high current rails, the device implements a low resistance N-channel MOSFET which reduces the drop out voltage across the device.

The device starts its operation by monitoring the VIN bus. When VIN exceeds the undervoltage-lockout threshold ( $V_{UVL}$ ), the device samples the EN/UVLO pin. A high level on this pin enables the internal MOSFET. When VIN rises, the internal MOSFET of the device starts conducting and allow current to flow from VIN to VOUT. When EN/UVLO is held low (below  $V_{ENF}$ ), internal MOSFET is turned off.

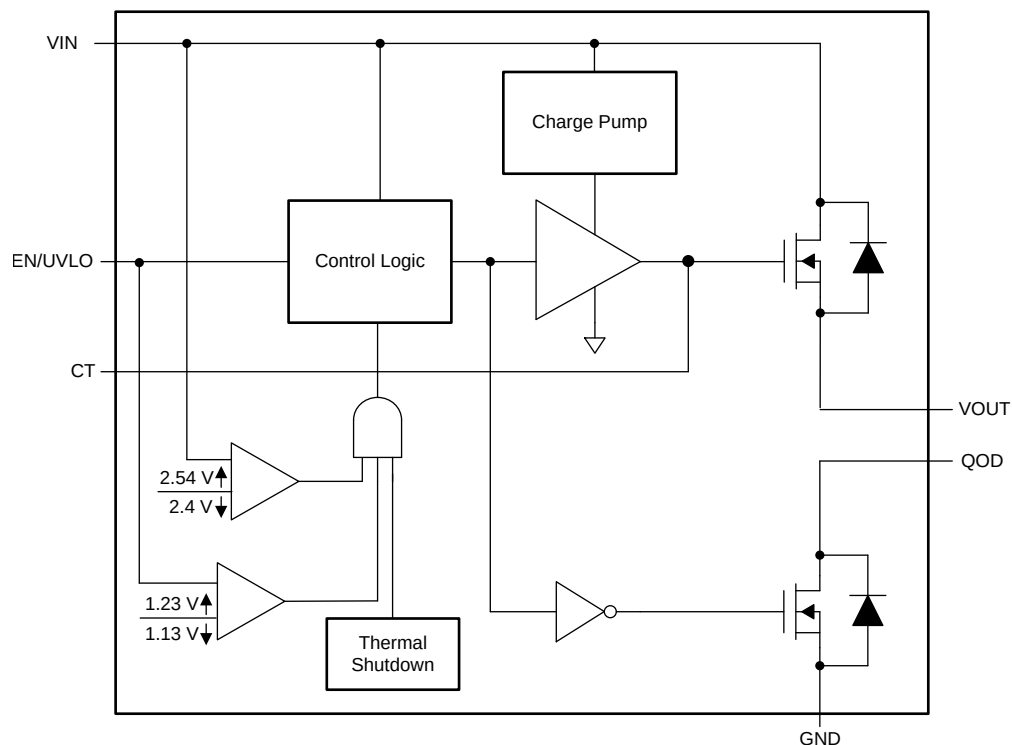
A voltage  $V_{EN/UVLO} < V_{ENF}$  on this pin turns off the internal FET, thus disconnecting VIN from VOUT, while voltage below  $V_{SHUTF}$  takes the device into shutdown mode, with  $I_Q$  less than 1  $\mu A$  to ensure minimal power loss.

The device has a configurable slew rate which helps reduce or eliminate power supply droop because of large inrush currents. The device also features a QOD (Quick Output Discharge) pin with an internal pull-down resistance ( $R_{PD}$ ) which can be used to discharge VOUT once the switch is disabled.

During shutdown, the device has very low leakage currents, thereby reducing unnecessary leakages for downstream modules during standby. Integrated control logic, driver, charge pump, and output discharge FET eliminates the need for any external components which reduces solution size and bill of materials (BOM) count.

The device has a thermal protection feature to protect itself against thermal damage due to overtemperature and overcurrent conditions. Safe Operating Area (SOA) requirements are thus inherently met without any special design consideration by the board designer.

### 8.2 Functional Block Diagram



## 8.3 Feature Description

### 8.3.1 On and Off Control

The EN/UVLO pin controls the state of the switch. EN/UVLO is active high and has a low threshold that can interface with low-voltage signals. The EN/UVLO pin is compatible with standard GPIO logic threshold. It can be used with any microcontroller with 1.2 V or higher GPIO voltage. This pin cannot be left floating and must be driven either high or low for proper functionality.

### 8.3.2 Quick Output Discharge (QOD)

The TPS22810-Q1 includes a QOD feature. The QOD pin can be configured in one of three ways:

- QOD pin shorted to VOUT pin. Using this method, the discharge rate after the switch becomes disabled is controlled with the value of the internal pull-down resistance ( $R_{PD}$ ). The value of this resistance is listed in the [Electrical Characteristics](#) table.
- QOD pin connected to VOUT pin using an external resistor  $R_{EXT}$ . After the switch becomes disabled, the discharge rate is controlled by the value of the total resistance of the QOD. To adjust the total QOD resistance, [公式 1](#) can be used.

$$R_{QOD} = R_{PD} + R_{EXT}$$

where

- $R_{QOD}$  is the total output discharge resistance
- $R_{PD}$  is the internal pulldown resistance
- $R_{EXT}$  is the external resistance placed between the VOUT and QOD pin. (1)
- QOD pin is unused and left floating. Using this method, there is no quick output discharge functionality, and the output remains floating after the switch is disabled.

Note that during thermal shutdown, the QOD functionality is not available. The device does not discharge the load because  $R_{PD}$  does not become engaged.

The fall times of the device depend on many factors including the total resistance of the QOD,  $V_{IN}$ , and the output capacitance. When QOD is connected to VOUT, the fall time changes over  $V_{IN}$  because the internal  $R_{PD}$  varies over  $V_{IN}$ . To calculate the approximate fall time of  $V_{OUT}$  for a given  $R_{QOD}$ , use [公式 2](#) and [表 1](#).

$$V_{CAP} = V_{IN} \times e^{-t/\tau}$$

where

- $V_{CAP}$  is the voltage across the capacitor (V)
- $t$  is the time since power supply removal (s)
- $\tau$  is the time constant equal to  $R_{QOD} \times C_L$  (2)

The fall time's dependency on  $V_{IN}$  becomes minimal because the QOD value increases with additional external resistance. See [表 1](#) for QOD fall times.

**表 1. QOD Fall Times**

| $V_{IN}$ (V) | FALL TIME ( $\mu$ s) 90% - 10%, $C_{IN} = 1 \mu$ F, $I_{OUT} = 0$ A, $V_{IN} = 0$ V, ON = 0 V <sup>(1)</sup> |                  |                   |                          |                  |                   |
|--------------|--------------------------------------------------------------------------------------------------------------|------------------|-------------------|--------------------------|------------------|-------------------|
|              | $T_A = 25^\circ\text{C}$                                                                                     |                  |                   | $T_A = 85^\circ\text{C}$ |                  |                   |
|              | $C_L = 1 \mu$ F                                                                                              | $C_L = 10 \mu$ F | $C_L = 100 \mu$ F | $C_L = 1 \mu$ F          | $C_L = 10 \mu$ F | $C_L = 100 \mu$ F |
| 18           | 470                                                                                                          | 4700             | 47000             | 470                      | 4700             | 47000             |
| 12           | 450                                                                                                          | 4500             | 45000             | 450                      | 4500             | 45000             |
| 9            | 440                                                                                                          | 4400             | 44000             | 440                      | 4400             | 44000             |
| 5            | 500                                                                                                          | 5000             | 50000             | 480                      | 4800             | 48000             |
| 3.3          | 600                                                                                                          | 6000             | 60000             | 570                      | 5700             | 57000             |

(1) TYPICAL VALUES WITH QOD SHORTED TO VOUT

### 8.3.2.1 QOD when System Power is Removed

The adjustable QOD can be used to control the power down sequencing of a system even when the system power supply is removed. When the power is removed, the input capacitor,  $C_{IN}$ , discharges at  $V_{IN}$ . Past the set UVLO level, the pull-down resistance  $R_{PD}$  becomes disabled and the output no longer becomes discharged. If there is still remaining charge on the output capacitor, this results in longer fall times. Care must be taken such that  $C_{IN}$  is large enough to meet the device UVLO settings.

### 8.3.2.2 Internal QOD Considerations

Special considerations must be taken when using the internal  $R_{PD}$  by shorting the QOD pin to the VOUT pin. The internal  $R_{PD}$  is a pull-down resistance designed to quickly discharge a load after the switch has been disabled. Care must be used to ensure that excessive current does not flow through  $R_{PD}$  during discharge so that the maximum  $T_J$  of 125°C is not exceeded. When using only the internal  $R_{PD}$  to discharge a load, the total capacitive load must not exceed 200  $\mu$ F. Otherwise, an external resistor,  $R_{EXT}$  must be used to ensure the amount of current flowing through  $R_{PD}$  is properly limited and the maximum  $T_J$  is not exceeded. To ensure the device is not damaged, the remaining charge from  $C_L$  must decay naturally through the internal QOD resistance and must not be driven.

### 8.3.3 EN/UVLO

EN/UVLO controls the ON and OFF state of the internal MOSFET, as an input pin. In its high state, the internal MOSFET is enabled. A low on this pin turns off the internal MOSFET. High and Low levels are specified in the parametric table of the datasheet.

A voltage  $V_{EN/UVLO} < V_{ENF}$  on this pin turns off the internal FET, thus disconnecting  $V_{IN}$  from VOUT, while voltage below  $V_{SHUTF}$  takes the device into shutdown mode, with  $I_Q$  less than 1  $\mu$ A to ensure minimal power loss.

The EN/UVLO pin can be directly driven by a 1.8 V, 3.3 V or 5 V general purpose output pin.

The internal de-glitch delay on EN/UVLO falling edge is intentionally kept low (2.5  $\mu$ s typical) for quick detection of power failure. For applications where a higher de-glitch delay on EN/UVLO is desired, or when the supply is particularly noisy, it is recommended to use an external bypass capacitor from EN/UVLO to GND.

The undervoltage lock out (UVLO) threshold can be programmed by using an external resistor divider from supply  $V_{IN}$  terminal to EN/UVLO terminal to GND shown in [Figure 18](#). When an undervoltage or input power fail event is detected, the internal FET is quickly turned off. If the programmable UVLO function is not needed, the EN/UVLO terminal must be connected to the  $V_{IN}$  terminal. EN/UVLO terminal must not be left floating.

The device also implements internal UVLO circuitry on the  $V_{IN}$  terminal. The device disables when the  $V_{IN}$  terminal voltage falls below internal UVLO Threshold ( $V_{UVF}$ ). The internal UVLO threshold has a hysteresis ( $V_{UVRhyst}$ ). See [Figure 19](#) and [Figure 20](#).

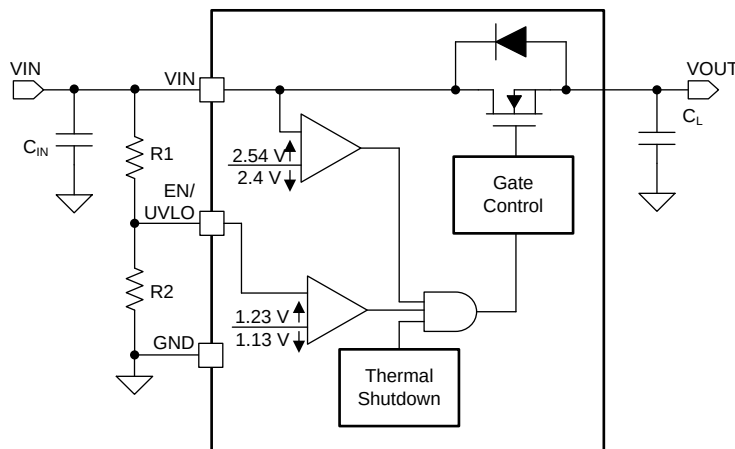


图 18. Configuring UVLO with External Resistor Network

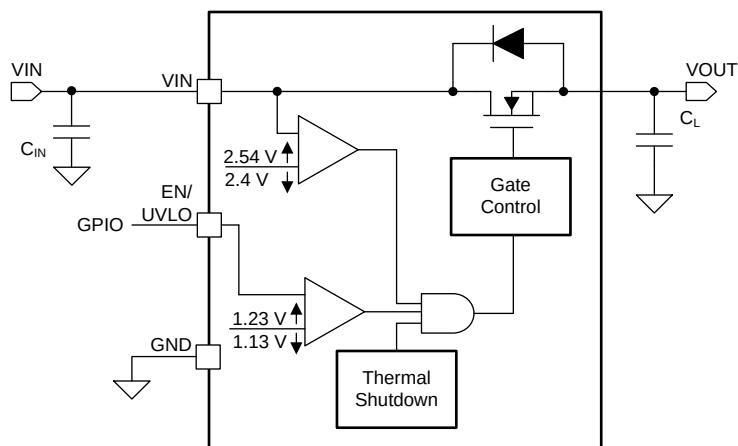


图 19. Using 1.8 V/3.3 V GPIO Signal Directly from Processor

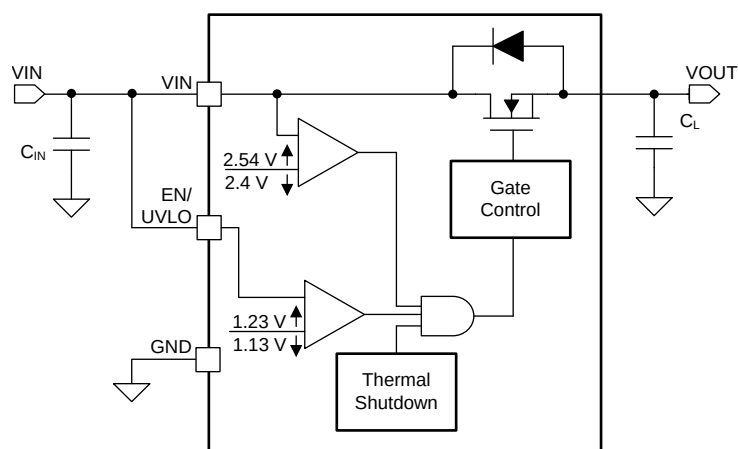


图 20. Default UVLO Threshold  $V_{UVLO}$  Using No Additional External Components

### 8.3.4 Adjustable Rise Time (CT)

A capacitor to GND on the CT pin sets the slew rate. The voltage on the CT pin can be as high as 2.5 V. An approximate formula for the relationship between CT and slew rate is shown in 公式 3. This equation accounts for 10% to 90% measurement on VOUT and does NOT apply for  $CT < 1 \text{ nF}$ .

Use 表 2 to determine rise times for when  $CT \geq 1 \text{ nF}$ .

$$SR = 46.62 / CT$$

where

- SR is the slew rate (in V/ $\mu\text{s}$ )
- CT is the the capacitance value on the CT pin (in pF)

Rise time can be calculated by dividing the input voltage by the slew rate. 表 2 describes rise time values measured on a typical device. Rise times shown below are only valid for the power-up sequence where VIN is already in steady state condition before the EN/UVLO pin is asserted high.

**表 2. Rise Time Table**

| CT (pF) | RISE TIME ( $\mu$ s) 10% - 90%, $C_L = 0.1 \mu$ F, $C_{IN} = 1 \mu$ F, $R_L = 10 \Omega$ |            |           |           |             |
|---------|------------------------------------------------------------------------------------------|------------|-----------|-----------|-------------|
|         | VIN = 18 V                                                                               | VIN = 12 V | VIN = 9 V | VIN = 5 V | VIN = 3.3 V |
| 0       | 115                                                                                      | 91         | 78        | 60        | 98          |
| 470     | 136                                                                                      | 94         | 80        | 63        | 98          |
| 1000    | 310                                                                                      | 209        | 158       | 91        | 102         |
| 2200    | 688                                                                                      | 464        | 345       | 198       | 135         |
| 4700    | 1430                                                                                     | 957        | 704       | 397       | 265         |
| 10000   | 3115                                                                                     | 2085       | 1540      | 864       | 550         |
| 27000   | 8230                                                                                     | 5460       | 4010      | 2245      | 1430        |

### 8.3.5 Thermal Shutdown

The switch disables when the junction temperature ( $T_J$ ) rises above the thermal shutdown threshold,  $T_{SD}$ . The switch re-enables once the temperature drops below the  $T_{SD} - T_{SD,HYS}$  value.

## 8.4 Device Functional Modes

The features of the TPS22810-Q1 depend on the operating mode. 表 3 summarizes the Device Functional Modes.

**表 3. Function Table**

| EN/UVLO | Device State |
|---------|--------------|
| L       | Disabled     |
| H       | Enabled      |

## 9 Application and Implementation

### 注

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

### 9.1 Application Information

This section highlights some of the design considerations when implementing this device in various applications. A PSPICE model for this device is also available in the product page of this device on [www.ti.com](http://www.ti.com) (See the Device Support section for more information).

#### 9.1.1 ON and OFF Control

The EN/UVLO pin controls the state of the switch. Asserting EN/UVLO high enables the switch. EN/UVLO is active high and has a low threshold that can interface with low-voltage signals. The EN/UVLO pin is compatible with standard GPIO logic thresholds. It can be used with any microcontroller with 1.2 V or higher GPIO voltage. This pin cannot be left floating and must be driven either high or low for proper functionality.

#### 9.1.2 Input Capacitor (Optional)

To limit the voltage drop on the input supply caused by transient inrush currents when the switch turns on into a discharged load capacitor, a capacitor must be placed between VIN and GND. A 1- $\mu$ F ceramic capacitor,  $C_{IN}$ , placed close to the pins, is usually sufficient. Higher values of  $C_{IN}$  can be used to further reduce the voltage drop during high current applications. When switching heavy loads, it is recommended to have an input capacitor about 10 times higher than the output capacitor to avoid excessive voltage drop.

#### 9.1.3 Output Capacitor (Optional)

Due to the integrated body diode in the NMOS switch, a  $C_{IN}$  greater than  $C_L$  is highly recommended. A  $C_L$  greater than  $C_{IN}$  can cause VOUT to exceed VIN when the system supply is removed. This can result in current flow through the body diode from VOUT to VIN. A  $C_{IN}$  to  $C_L$  ratio of 10 to 1 is recommended for minimizing VIN dip caused by inrush currents during startup; however, a 10 to 1 ratio for capacitance is not required for proper functionality of the device. A ratio smaller than 10 to 1 (such as 1 to 1) can cause slightly more VIN dip upon turnon due to inrush currents.

This can be mitigated by increasing the capacitance on the CT pin for a longer rise time.

### 9.2 Typical Application

This typical application demonstrates how the TPS22810-Q1 can be used to power downstream modules.

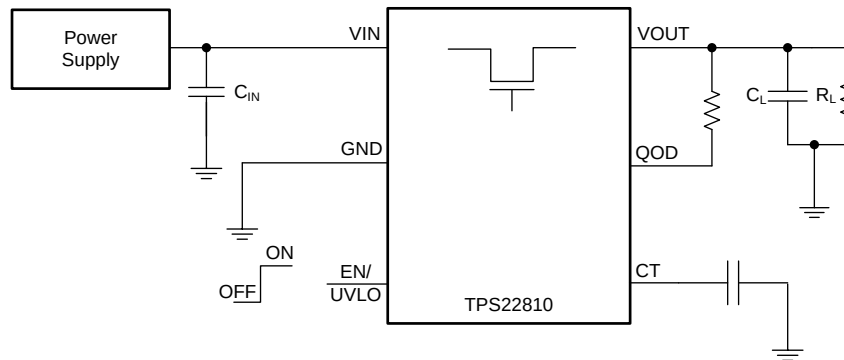


图 21. Typical Application Schematic



## Typical Application (接下页)

### 9.2.1 Design Requirements

For this design example, use the values listed in 表 4:

**表 4. Design Parameters**

| DESIGN PARAMETER                  | EXAMPLE VALUE |
|-----------------------------------|---------------|
| V <sub>IN</sub>                   | 12 V          |
| Load current                      | 2 A           |
| C <sub>L</sub>                    | 22 μF         |
| Desired fall time                 | 20 ms         |
| Maximum acceptable inrush current | 400 mA        |

### 9.2.2 Detailed Design Procedure

#### 9.2.2.1 Shutdown Sequencing During Unexpected Power Loss

Using the adjustable Quick Output Discharge function of the TPS22810-Q1, adding a load switch to each power rail can be used to manage the power down sequencing in the event of an unexpected power loss (for example, battery removal). To determine the QOD values for each load switch, first confirm the power down order of the device you wish to power sequence. Be sure to check if there are voltage or timing margins that must be maintained during power down. Next, consult 表 1 to determine appropriate C<sub>L</sub> and R<sub>QOD</sub> values for each power rail's load switch so that the load switches' fall times correspond to the order in which they need to be powered down. In the above example, we must have this power rail's fall time to be 4 ms. Using 公式 2, we can determine the appropriate R<sub>QOD</sub> to achieve our desired fall time.

Since fall times are measured from 90% of V<sub>OUT</sub> to 10% of V<sub>OUT</sub>, using 公式 2, we get 公式 4 and 公式 5.

$$1.2\text{V} = 10.8\text{V} \times e^{-(20\text{ms})/(R_{\text{QOD}} \times (22\mu\text{F}))} \quad (4)$$

$$R_{\text{QOD}} = 413.7 \, \Omega \quad (5)$$

Consulting 图 6, R<sub>PD</sub> at V<sub>IN</sub> = 12 V is approximately 250 Ω. Using 公式 1, the required external QOD resistance can be calculated shown in 公式 6 and 公式 7.

$$413.7 \, \Omega = 250 \, \Omega + R_{\text{EXT}} \quad (6)$$

$$R_{\text{EXT}} = 163.7 \, \Omega \quad (7)$$

图 22 through 图 25 are scope shots demonstrating an example of the QOD functionality when power is removed from the device (both ON and VIN are disconnected simultaneously). In the scope shots, the V<sub>IN</sub> = 12 V and correspond to when R<sub>QOD</sub> = 1000 Ω, R<sub>QOD</sub> = 500 Ω, and QOD = V<sub>OUT</sub> with two values of C<sub>L</sub> = 10 μF and 22 μF.

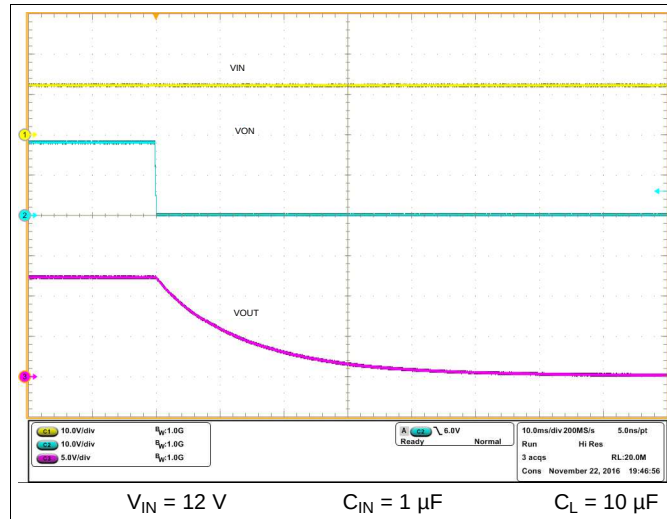


图 22. Fall Time  $t_F$  at  $V_{IN} = 12\text{ V}$ ,  $R_{QOD} = 1000\text{ }\Omega$

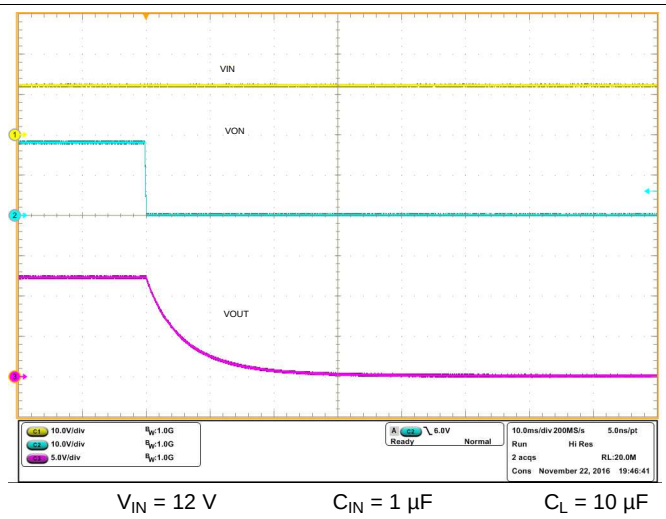


图 23. Fall Time  $t_F$  at  $V_{IN} = 12\text{ V}$ ,  $R_{QOD} = 500\text{ }\Omega$

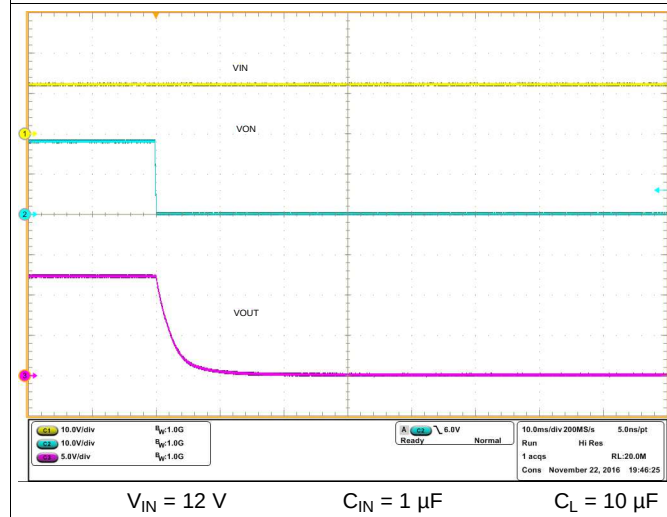


图 24.  $t_F$  at  $V_{IN} = 12\text{ V}$ ,  $Q_{OD} = V_{OUT}$

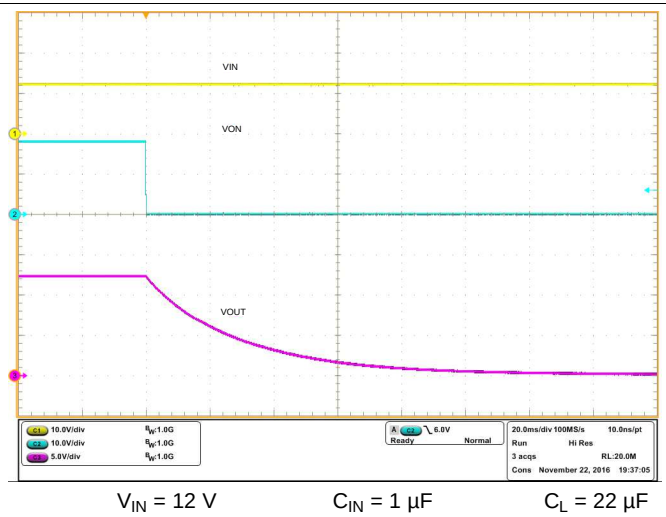
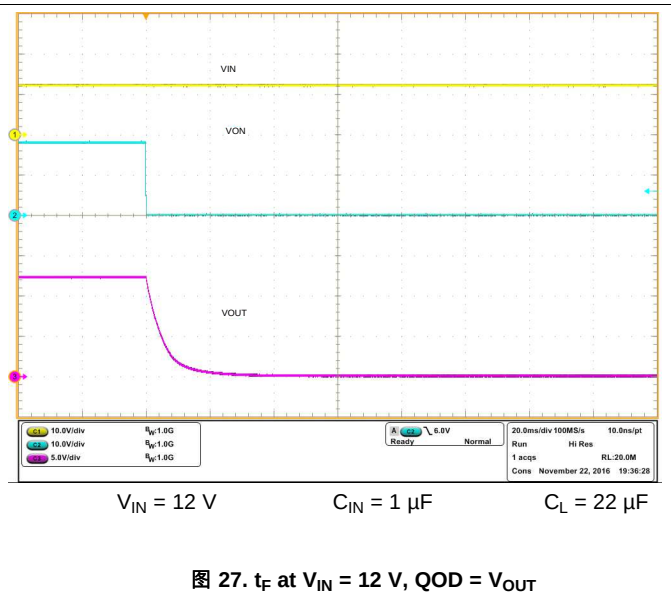
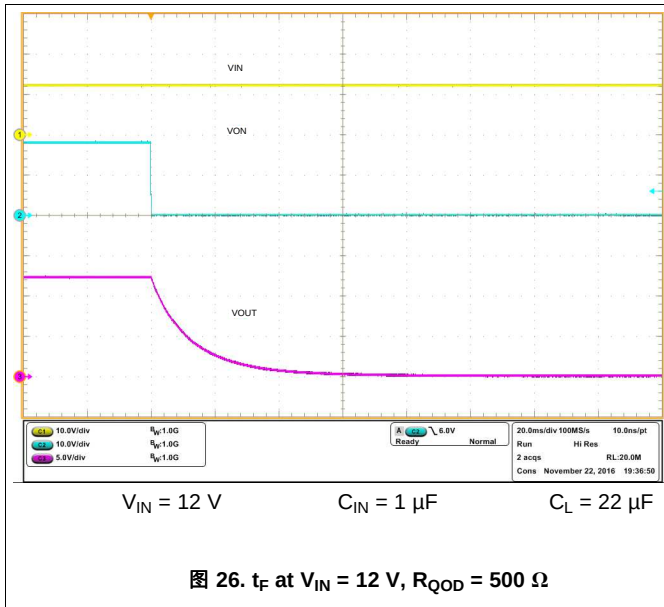


图 25.  $t_F$  at  $V_{IN} = 12\text{ V}$ ,  $R_{QOD} = 1000\text{ }\Omega$



### 9.2.2.2 VIN to VOUT Voltage Drop

The  $V_{IN}$  to  $V_{OUT}$  voltage drop in the device is determined by the  $R_{ON}$  of the device and the load current. The  $R_{ON}$  of the device depends upon the  $V_{IN}$  conditions of the device. Refer to the  $R_{ON}$  specification of the device in the [Electrical Characteristics](#) table of this datasheet. Once the  $R_{ON}$  of the device is determined based upon the  $V_{IN}$  conditions, use [公式 8](#) to calculate the  $V_{IN}$  to  $V_{OUT}$  voltage drop.

$$\Delta V = I_{LOAD} \times R_{ON}$$

where

- $\Delta V$  is the voltage drop from  $V_{IN}$  to  $V_{OUT}$
- $I_{LOAD}$  is the load current
- $R_{ON}$  is the On-resistance of the device for a specific  $V_{IN}$

(8)

An appropriate  $I_{LOAD}$  must be chosen such that the  $I_{MAX}$  specification of the device is not violated.

### 9.2.2.3 Inrush Current

To determine how much inrush current is caused by the  $C_L$  capacitor, use [公式 9](#).

$$I_{INRUSH} = C_L \times \frac{dV_{OUT}}{dt}$$

where

- $I_{INRUSH}$  is the amount of inrush caused by  $C_L$
- $C_L$  is the capacitance on  $V_{OUT}$
- $dt$  is the Output Voltage rise time during the ramp up of  $V_{OUT}$  when the device is enabled
- $dV_{OUT}$  is the change in  $V_{OUT}$  during the ramp up of  $V_{OUT}$  when the device is enabled

(9)

The appropriate rise time can be calculated using the design requirements and the inrush current equation. When we calculate the rise time (measured from 10% to 90% of  $V_{OUT}$ ), we account for this in our  $dV_{OUT}$  parameter (80% of  $V_{OUT} = 9.6\text{ V}$ ) shown in [公式 10](#) and [公式 11](#).

$$400\text{ mA} = 22\ \mu\text{F} \times 9.6\text{ V}/dt \quad (10)$$

$$dt = 528\ \mu\text{s} \quad (11)$$

To ensure an inrush current of less than 400 mA, choose a CT value that yields a rise time of more than 528  $\mu\text{s}$ . Consulting [表 2](#) at  $V_{IN} = 12\text{ V}$ ,  $CT = 4700\text{ pF}$  provides a typical rise time of 957  $\mu\text{s}$ . Using this rise time and voltage into [公式 9](#), yields [公式 12](#) and [公式 13](#).

$$I_{Inrush} = 22\ \mu\text{F} \times 9.6\text{ V}/957\ \mu\text{s} \quad (12)$$

$$I_{Inrush} = 220\text{ mA} \quad (13)$$

An appropriate  $C_L$  value must be placed on VOUT such that the  $I_{MAX}$  and  $I_{PLS}$  specifications of the device are not violated.

### 9.2.3 Application Curves

See the oscilloscope captures below for an example of how the CT capacitor can be used to reduce inrush current for  $V_{IN} = 12\text{ V}$ . See the [Adjustable Rise Time \(CT\)](#) section for rise times for corresponding CT values.

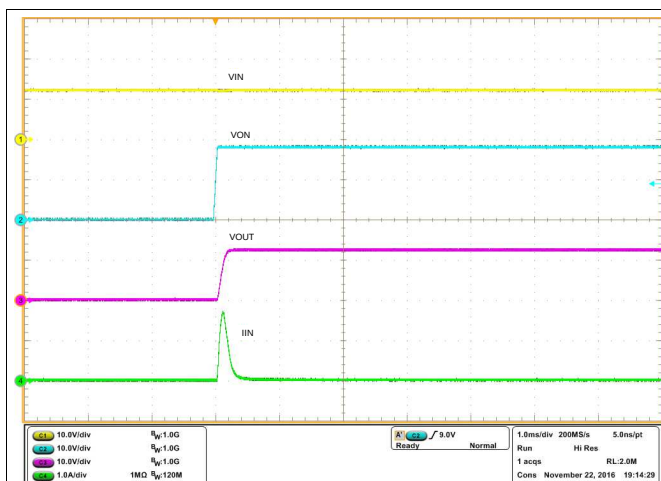


图 28. TPS22810-Q1 Inrush Current With  $C_L = 22\text{ }\mu\text{F}$ ,  $C_T = 0\text{ pF}$

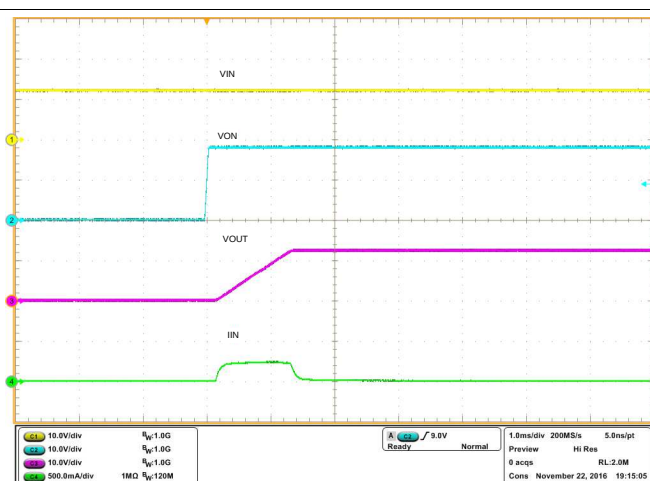


图 29. TPS22810-Q1 Inrush Current with  $C_L = 22\text{ }\mu\text{F}$ ,  $C_T = 4700\text{ pF}$

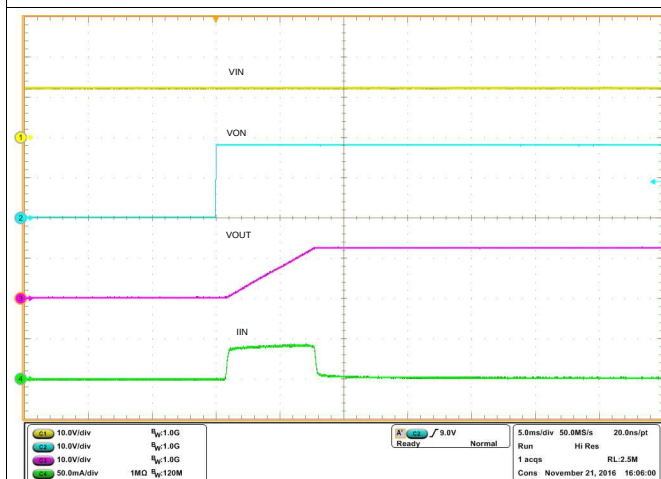


图 30. TPS22810-Q1 Inrush Current With  $C_L = 22\text{ }\mu\text{F}$ ,  $C_T = 27000\text{ pF}$

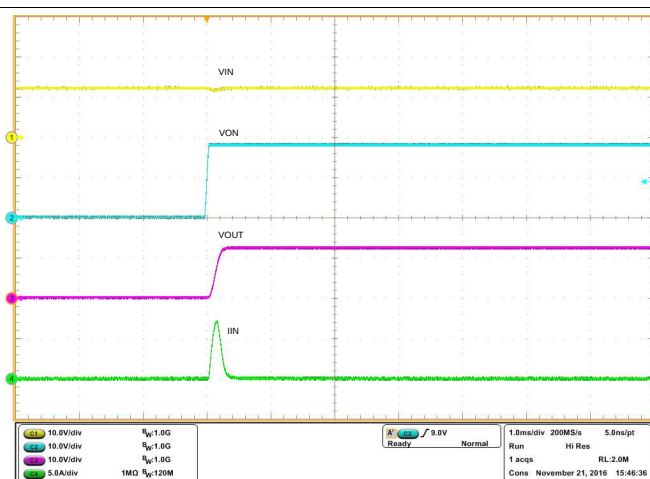
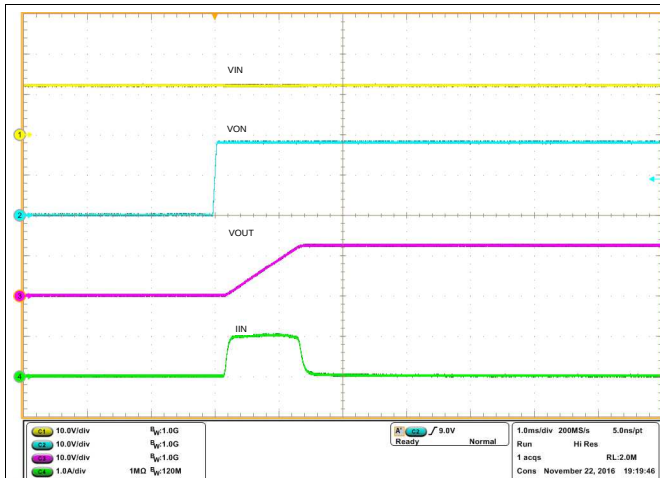
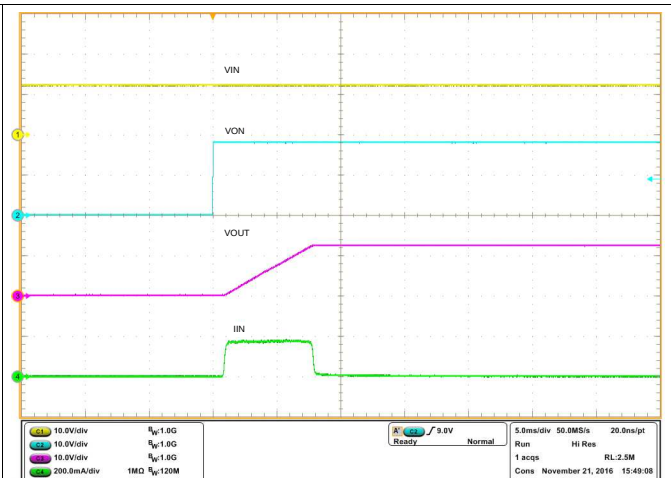


图 31. TPS22810-Q1 Inrush Current With  $C_L = 100\text{ }\mu\text{F}$ ,  $C_T = 0\text{ pF}$



**图 32. TPS22810-Q1 Inrush Current**  
With  $C_L = 100 \mu\text{F}$ ,  $C_T = 4700 \text{ pF}$



**图 33. TPS22810-Q1 Inrush Current**  
With  $C_L = 100 \mu\text{F}$ ,  $C_T = 27000 \text{ pF}$

## 10 Power Supply Recommendations

The device is designed to operate from a VIN range of 2.7 V to 18 V. This supply must be well regulated and placed as close to the device terminal as possible with the recommended 1- $\mu\text{F}$  bypass capacitor. If the supply is located more than a few inches from the device terminals, additional bulk capacitance may be required in addition to the ceramic bypass capacitors. If additional bulk capacitance is required, an electrolytic, tantalum, or ceramic capacitor of 1- $\mu\text{F}$  may be sufficient.

The TPS22810-Q1 operates regardless of power sequencing order. The order in which voltages are applied to VIN and EN/UVLO does not damage the device as long as the voltages do not exceed the absolute maximum operating conditions.

## 11 Layout

### 11.1 Layout Guidelines

1. VIN and VOUT traces must be as short and wide as possible to accommodate for high current.
2. The VIN pin must be bypassed to ground with low ESR ceramic bypass capacitors. The typical recommended bypass capacitance is 1-μF ceramic with X5R or X7R dielectric. This capacitor must be placed as close to the device pins as possible.

### 11.2 Layout Example

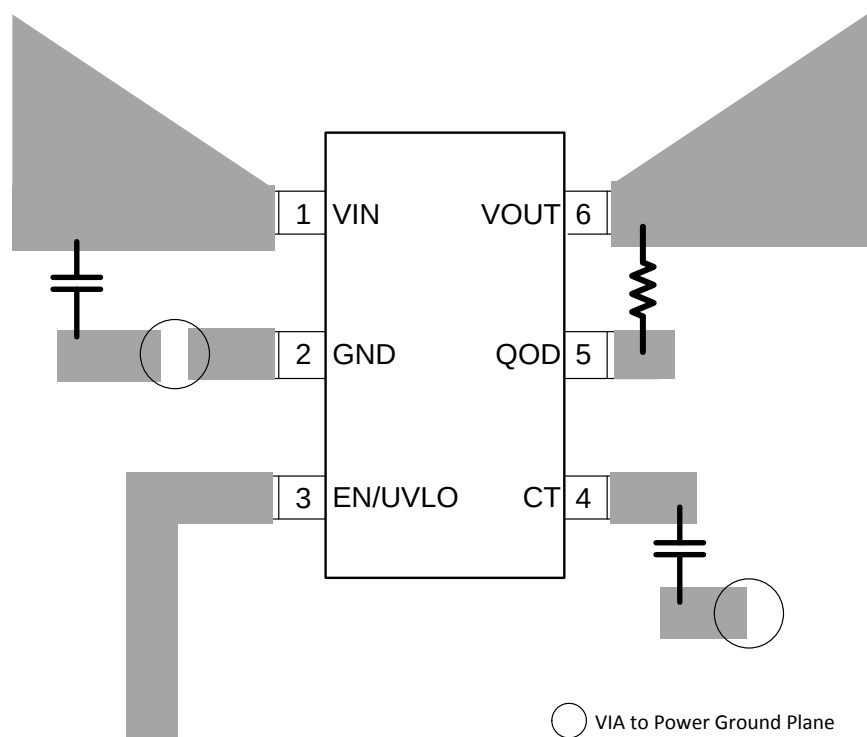


图 34. Recommended Board Layout

### 11.3 Thermal Considerations

For best performance, all traces must be as short as possible. To be most effective, the input and output capacitors must be placed close to the device to minimize the effects that parasitic trace inductances may have on normal and short-circuit operation. Using wide traces for VIN, VOUT, and GND helps minimize the parasitic electrical effects along with minimizing the case to ambient thermal impedance.

The maximum IC junction temperature must be restricted to 150°C under normal operating conditions. To calculate the maximum allowable dissipation,  $P_{D(max)}$  for a given output current and ambient temperature, use 公式 14.

$$P_{D(MAX)} = \frac{T_{J(MAX)} - T_A}{\theta_{JA}}$$

where

- $P_{D(MAX)}$  is the maximum allowable power dissipation
- $T_{J(MAX)}$  is the maximum allowable junction temperature (150°C for the TPS22810-Q1)
- $T_A$  is the ambient temperature of the device
- $\theta_{JA}$  is the junction to air thermal impedance. Refer to the [Thermal Information](#) table. This parameter highly depends on the board layout.

(14)

## 12 器件和文档支持

### 12.1 器件支持

#### 12.1.1 开发支持

关于 TPS22810 PSpice 瞬态模型，请参见 [《TPS22810 PSpice 瞬态模型》](#)

### 12.2 文档支持

#### 12.2.1 相关文档

请参阅如下相关文档：

- [TPS22810 负载开关评估模块](#)
- [选择一个负载开关以代替分立式解决方案](#)
- [负载开关的计时](#)

### 12.3 接收文档更新通知

要接收文档更新通知（包括芯片勘误表），请转至 [ti.com.cn](http://ti.com.cn) 上您的器件对应的产品文件夹。单击右上角的通知我按钮。点击后，您将每周定期收到已更改的产品信息（如果有的话）。有关更改的详细信息，请查看任意已修订文档的修订历史记录。

### 12.4 社区资源

下列链接提供到 TI 社区资源的连接。链接的内容由各个分销商“按照原样”提供。这些内容并不构成 TI 技术规范，并且不一定反映 TI 的观点；请参阅 TI 的 [《使用条款》](#)。

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**设计支持** **TI 参考设计支持** 可帮助您快速查找有帮助的 E2E 论坛、设计支持工具以及技术支持的联系信息。

### 12.5 商标

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### 12.6 静电放电警告



ESD 可能会损坏该集成电路。德州仪器 (TI) 建议通过适当的预防措施处理所有集成电路。如果不遵守正确的处理措施和安装程序，可能会损坏集成电路。

ESD 的损坏小至导致微小的性能降级，大至整个器件故障。精密的集成电路可能更容易受到损坏，这是因为非常细微的参数更改都可能会导致器件与其发布的规格不相符。

### 12.7 Glossary

[SLYZ022](#) — *TI Glossary*.

This glossary lists and explains terms, acronyms, and definitions.

## 13 机械、封装和可订购信息

以下页面包含机械、封装和可订购信息。这些信息是指定器件的最新可用数据。数据如有变更，恕不另行通知，且不会对此文档进行修订。如需获取此数据表的浏览器版本，请参阅左侧的导航栏。

## PACKAGING INFORMATION

| Orderable Device | Status<br>(1) | Package Type | Package<br>Drawing | Pins | Package<br>Qty | Eco Plan<br>(2) | Lead finish/<br>Ball material<br>(6) | MSL Peak Temp<br>(3) | Op Temp (°C) | Device Marking<br>(4/5) | Samples                 |
|------------------|---------------|--------------|--------------------|------|----------------|-----------------|--------------------------------------|----------------------|--------------|-------------------------|-------------------------|
| TPS22810TDBVRQ1  | ACTIVE        | SOT-23       | DBV                | 6    | 3000           | RoHS & Green    | NIPDAU                               | Level-2-260C-1 YEAR  | -40 to 105   | 1EFF                    | <a href="#">Samples</a> |

(1) The marketing status values are defined as follows:

**ACTIVE:** Product device recommended for new designs.

**LIFEBUY:** TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

**NRND:** Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) **RoHS:** TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

**RoHS Exempt:** TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

**Green:** TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

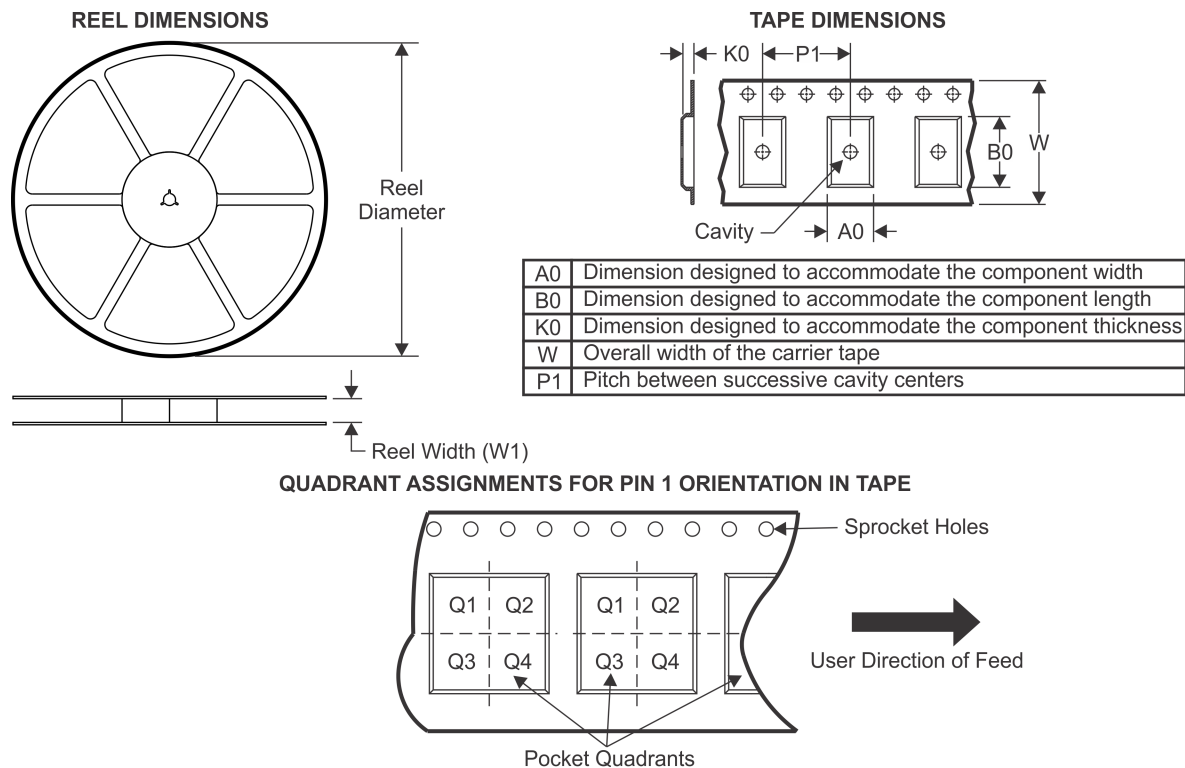
(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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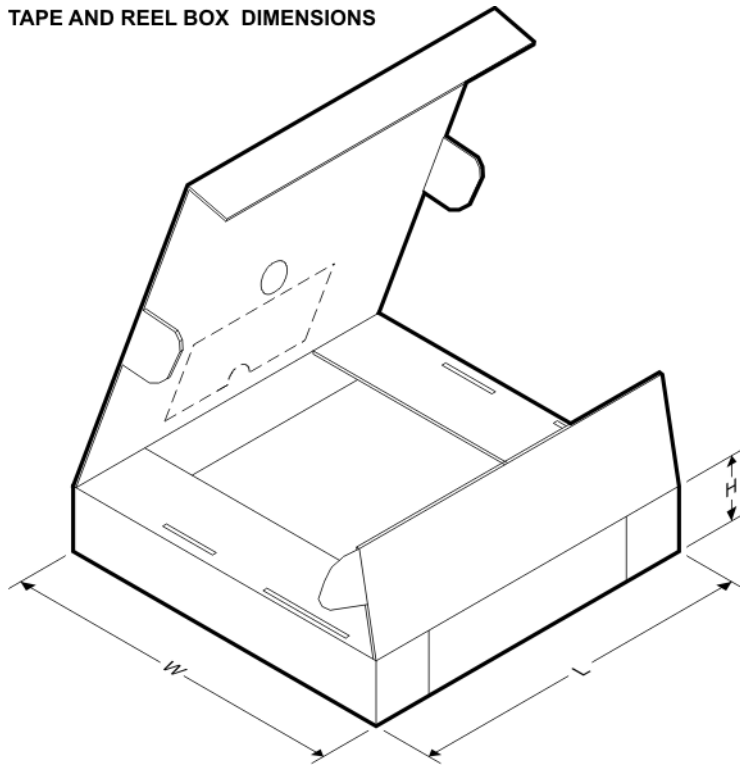


**TAPE AND REEL INFORMATION**


\*All dimensions are nominal

| Device          | Package Type | Package Drawing | Pins | SPQ  | Reel Diameter (mm) | Reel Width W1 (mm) | A0 (mm) | B0 (mm) | K0 (mm) | P1 (mm) | W (mm) | Pin1 Quadrant |
|-----------------|--------------|-----------------|------|------|--------------------|--------------------|---------|---------|---------|---------|--------|---------------|
| TPS22810TDBVRQ1 | SOT-23       | DBV             | 6    | 3000 | 180.0              | 8.4                | 3.2     | 3.2     | 1.4     | 4.0     | 8.0    | Q3            |

## TAPE AND REEL BOX DIMENSIONS



\*All dimensions are nominal

| Device          | Package Type | Package Drawing | Pins | SPQ  | Length (mm) | Width (mm) | Height (mm) |
|-----------------|--------------|-----------------|------|------|-------------|------------|-------------|
| TPS22810TDBVRQ1 | SOT-23       | DBV             | 6    | 3000 | 210.0       | 185.0      | 35.0        |

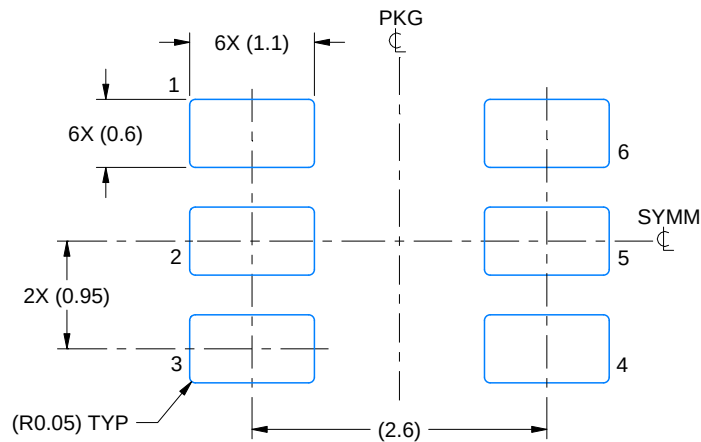


# EXAMPLE BOARD LAYOUT

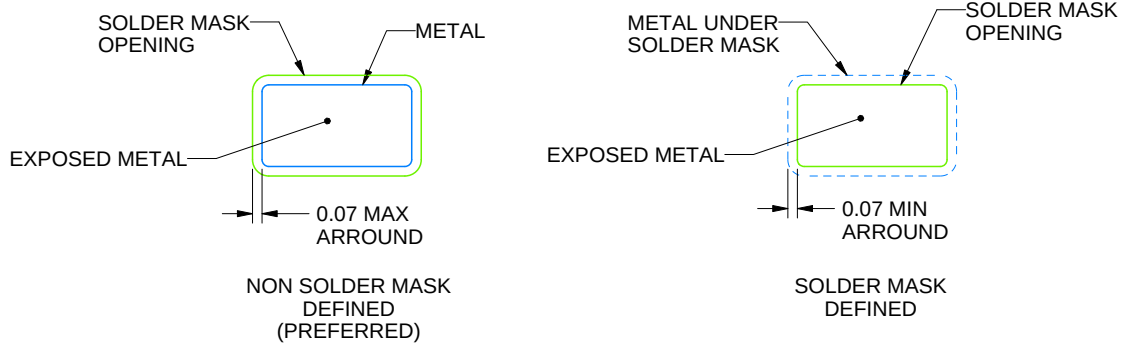
DBV0006A

SOT-23 - 1.45 mm max height

SMALL OUTLINE TRANSISTOR



LAND PATTERN EXAMPLE  
EXPOSED METAL SHOWN  
SCALE:15X



SOLDER MASK DETAILS

4214840/B 03/2018

NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

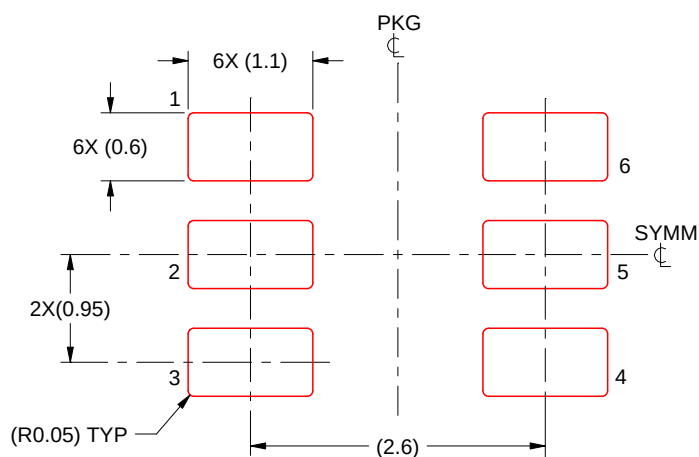
7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

## EXAMPLE STENCIL DESIGN

DBV0006A

SOT-23 - 1.45 mm max height

SMALL OUTLINE TRANSISTOR



SOLDER PASTE EXAMPLE  
BASED ON 0.125 mm THICK STENCIL  
SCALE:15X

4214840/B 03/2018

NOTES: (continued)

8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
9. Board assembly site may have different recommendations for stencil design.

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